

# < DIPIPM > PSS20S71F6

TRANSFER MOLDING TYPE INSULATED TYPE

#### **OUTLINE**



#### MAIN FUNCTION AND RATINGS

- 3 phase DC/AC inverter
- 600V / 20A (CSTBT)
- N-side IGBT open emitter
- Built-in bootstrap diodes with current limiting resistor

#### **APPLICATION**

 AC 100~240Vrms(DC voltage:400V or below) class low power motor control

#### TYPE NAME

PSS20S71F6 With temperature output function

#### INTEGRATED DRIVE, PROTECTION AND SYSTEM CONTROL FUNCTIONS

• For P-side : Drive circuit, High voltage high-speed level shifting, Control supply under-voltage (UV) protection

• For N-side : Drive circuit, Control supply under-voltage protection (UV), Short circuit protection (SC),

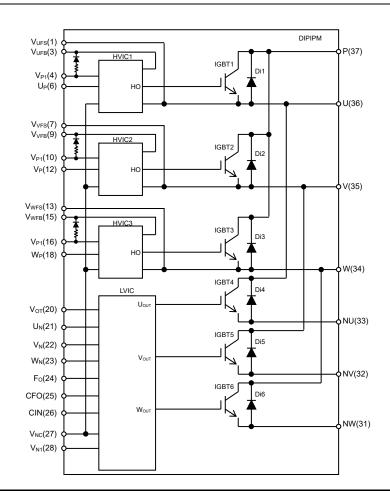
• Fault signaling : Corresponding to SC fault (N-side IGBT), UV fault (N-side supply)

• Temperature output : Outputting LVIC temperature by analog signal

• Input interface : 3, 5V line, Schmitt trigger receiver circuit (High Active)

• UL Recognized : UL1557 File E80276

#### **INTERNAL CIRCUIT**



Publication Date: January 2021

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#### **MAXIMUM RATINGS** (T<sub>j</sub> = 25°C, unless otherwise noted)

#### **INVERTER PART**

Symbol	Parameter	Condition		Ratings	Unit
V <sub>CC</sub>	Supply voltage	Applied between P-NU,NV,NW		450	V
V <sub>CC(surge)</sub>	Supply voltage (surge)	Applied between P-NU,NV,NW		500	V
V <sub>CES</sub>	Collector-emitter voltage			600	V
±l <sub>C</sub>	Each IGBT collector current	$T_C = 25^{\circ}C$ (Not	te 1)	20	Α
±I <sub>CP</sub>	Each IGBT collector current (peak)	T <sub>C</sub> = 25°C, less than 1ms		40	Α
Pc	Collector dissipation	T <sub>C</sub> = 25°C, per 1 chip		76.9	W
Tj	Junction temperature			-20~+150	°C

Note1: Pulse width and period are limited due to junction temperature.

#### **CONTROL (PROTECTION) PART**

Symbol	Parameter	Condition	Ratings	Unit
V <sub>D</sub>	Control supply voltage	Applied between V <sub>P1</sub> -V <sub>NC</sub> , V <sub>N1</sub> -V <sub>NC</sub>	20	V
$V_{DB}$	Control supply voltage	Applied between V <sub>UFB</sub> -V <sub>UFS</sub> , V <sub>VFB</sub> -V <sub>VFS</sub> , V <sub>WFB</sub> -V <sub>WFS</sub>	20	V
V <sub>IN</sub>	Input voltage	Applied between U <sub>P</sub> , V <sub>P</sub> , W <sub>P</sub> -V <sub>PC</sub> , U <sub>N</sub> , V <sub>N</sub> , W <sub>N</sub> -V <sub>NC</sub>	-0.5~V <sub>D</sub> +0.5	V
$V_{FO}$	Fault output supply voltage	Applied between F <sub>O</sub> -V <sub>NC</sub>	-0.5~V <sub>D</sub> +0.5	V
I <sub>FO</sub>	Fault output current	Sink current at F <sub>0</sub> terminal	1	mA
V <sub>SC</sub>	Current sensing input voltage	Applied between CIN-V <sub>NC</sub>	-0.5~V <sub>D</sub> +0.5	V

#### **TOTAL SYSTEM**

Symbol	Parameter	Condition	Ratings	Unit
V <sub>CC(PROT)</sub>	Self protection supply voltage limit (Short circuit protection capability)	V <sub>D</sub> = 13.5~16.5V, Inverter Part T <sub>i</sub> = 125°C, non-repetitive, less than 2μs	400	V
T <sub>C</sub>	Module case operation temperature	nperature Measurement point of Tc is provided in Fig.1		°C
T <sub>stg</sub>	Storage temperature		-40~+125	°C
V <sub>iso</sub>	Isolation voltage	60Hz, Sinusoidal, AC 1min, between connected all pins and heat sink plate	2500	V <sub>rms</sub>

Fig. 1: T<sub>C</sub> MEASUREMENT POINT

# IGBT chip position Power terminals Control terminals Groove Heat sink side

#### THERMAL RESISTANCE

Symbol	Doromotor	Condition		Limits		
Symbol Parameter		Condition		Тур.	Max.	Unit
R <sub>th(j-c)Q</sub>	Junction to case thermal	Inverter IGBT part (per 1/6 module)		-	1.3	K/W
R <sub>th(j-c)F</sub>	resistance (Note 2)	Inverter FWDi part (per 1/6 module)	-	-	3.0	K/W

Note 2: Grease with good thermal conductivity and long-term endurance should be applied evenly with about +100µm~+200µm on the contacting surface of DIPIPM and heat sink. The contacting thermal resistance between DIPIPM case and heat sink Rth(c-f) is determined by the thickness and the thermal conductivity of the applied grease. For reference, Rth(c-f) is about 0.3K/W (per 1/6 module, grease thickness: 20µm, thermal conductivity: 1.0W/m•k).

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# **ELECTRICAL CHARACTERISTICS** ( $T_j = 25^{\circ}C$ , unless otherwise noted) **INVERTER PART**

Ch. a.l	Danamatan	0.000	Condition		Limits			Limit	
Symbol	Parameter	Condition			Min.	Тур.	Max.	Unit	
\/	Collector-emitter saturation	$V_D = V_{DB} = 15V, V_{IN} = 5V$	I <sub>C</sub> = 20A, T <sub>j</sub> = 25°C		-	1.40	1.90	V	
$V_{CE(sat)}$	voltage	V <sub>D</sub> -V <sub>DB</sub> - 13V, V <sub>IN</sub> - 3V	I <sub>C</sub> = 20A, T <sub>j</sub> = 125°C		-	1.50	2.00	v	
V <sub>EC</sub>	FWDi forward voltage	V <sub>IN</sub> = 0V, -I <sub>C</sub> = 20A			-	1.50	2.00	V	
t <sub>on</sub>					0.95	1.55	2.15	μs	
$t_{C(on)}$		$V_{CC}$ = 300V, $V_{D}$ = $V_{DB}$ = 15V			-	0.50	0.80	μs	
t <sub>off</sub>	Switching times	I <sub>C</sub> = 20A, T <sub>i</sub> = 125°C, V <sub>IN</sub> = 0↔5V			-	1.75	2.35	μs	
$t_{C(off)}$		Inductive Load (upper-lower a	Inductive Load (upper-lower arm)		-	0.40	0.60	μs	
t <sub>rr</sub>					-	0.30	-	μs	
1	Collector-emitter cut-off	\/ -\/	T <sub>j</sub> = 25°C		-	-	1	mΛ	
I <sub>CES</sub>	current	$V_{CE}=V_{CES}$ $T_i=125^{\circ}C$		-	-	10	mA		

**CONTROL (PROTECTION) PART** 

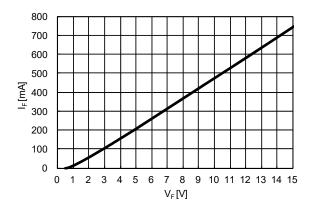
Symbol	Parameter	Cond	ition		Limits		Unit
Syllibol	Farameter	Cond	Ition	Min.	Тур.	Max.	Offic
		V <sub>D</sub> =15V, V <sub>IN</sub> =0V		-	-	6.00	
I <sub>D</sub>	Circuit current	Total of V <sub>P1</sub> -V <sub>NC</sub> , V <sub>N1</sub> -V <sub>NC</sub>	V <sub>D</sub> =15V, V <sub>IN</sub> =5V	-	-	6.00	mA
		Each part of V <sub>UFB</sub> - V <sub>UFS</sub> ,	$V_D=V_{DB}=15V$ , $V_{IN}=0V$	-	1	0.55	IIIA
I <sub>DB</sub>		$V_{VFB}$ - $V_{VFS}$ , $V_{WFB}$ - $V_{WFS}$	$V_D=V_{DB}=15V$ , $V_{IN}=5V$	-	-	0.55	
V <sub>SC(ref)</sub>	Short circuit trip level	V <sub>D</sub> = 15V	(Note 3)	0.45	0.48	0.51	V
$UV_DBt$	P-side Control supply		Trip level	10.0	-	12.0	V
$UV_DBr$	under-voltage protection(UV)	T <105°C	Reset level	10.5	-	12.5	V
$UV_Dt$	N-side Control supply	T <sub>j</sub> ≤125°C	Trip level	10.3	-	12.5	V
$UV_Dr$	under-voltage protection(UV)		Reset level	10.8	-	13.0	V
V <sub>OT</sub>	Temperature Output	Pull down R=5kΩ (Note 4)	LVIC Temperature=85°C	2.51	2.64	2.76	V
$V_{FOH}$	Fault output valtage	$V_{SC}$ = 0V, $F_O$ terminal pulled up	$V_{SC}$ = 0V, $F_O$ terminal pulled up to 5V by $10k\Omega$		-	-	V
$V_{FOL}$	Fault output voltage	V <sub>SC</sub> = 1V, I <sub>FO</sub> = 1mA		-	-	0.95	V
t <sub>FO</sub>	Fault output pulse width	C <sub>FO</sub> =22nF	(Note 5)	1.6	2.4	-	ms
I <sub>IN</sub>	Input current	V <sub>IN</sub> = 5V		0.70	1.00	1.50	mA
$V_{th(on)}$	ON threshold voltage			-	2.10	2.60	
$V_{th(off)}$	OFF threshold voltage	Applied between U <sub>P</sub> , V <sub>P</sub> , W <sub>P</sub> , l	0.80	1.30	-	V	
$V_{\text{th(hys)}}$	ON/OFF threshold hysteresis voltage	The section of the se	0.35	0.80	-	J	
$V_{F}$	Bootstrap Di forward voltage	I <sub>F</sub> =10mA including voltage drop by limiting resistor (Note 6)		0.5	0.9	1.3	V
R	Built-in limiting resistance	Included in bootstrap Di		16	20	24	Ω

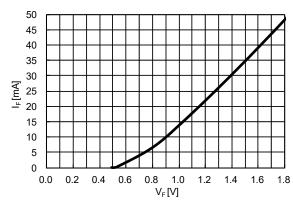
Note 3 : SC protection works only for N-side IGBT. Please select the external shunt resistance such that the SC trip-level is less than 2.0 times of the current rating.

4 : DIPIPM don't shutdown IGBTs and output fault signal automatically when temperature rises excessively. When temperature exceeds the protective level that user defined, controller (MCU) should stop the DIPIPM. Temperature of LVIC vs. VOT output characteristics is described in Fig. 3.

6 : The characteristics of bootstrap Di is described in Fig.2.

Fig. 2 Characteristics of bootstrap Di V<sub>F</sub>-I<sub>F</sub> curve (@Ta=25°C) including voltage drop by limiting resistor (Right chart is enlarged chart.)





<sup>5 :</sup> Fault signal Fo outputs when SC or UV protection works. Fo pulse width is different for each protection modes. At SC failure, Fo pulse width is a fixed width which is specified by the capacitor connected to C<sub>FO</sub> terminal. (C<sub>FO</sub>=9.1 x 10<sup>-6</sup> x t<sub>FO</sub> [F]), but at UV failure, Fo outputs continuously until recovering from UV state. (But minimum Fo pulse width is the specified time by C<sub>FO</sub>.)

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Fig. 3 Temperature of LVIC vs. VoT output characteristics

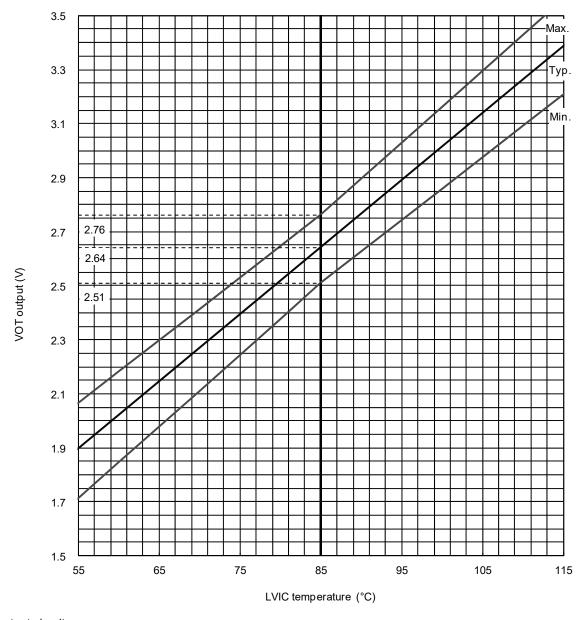
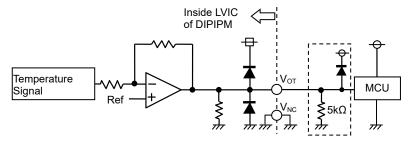


Fig. 4 VoT output circuit



- (1) It is recommended to insert  $5k\Omega$  (5.1 $k\Omega$  is recommended) pull down resistor for getting linear output characteristics at low temperature below room temperature. When the pull down resistor is inserted between  $V_{OT}$  and  $V_{NC}$ (control GND), the extra circuit current, which is calculated approximately by  $V_{OT}$  output voltage divided by pull down resistance, flows as LVIC circuit current continuously. In the case of using  $V_{OT}$  for detecting high temperature over room temperature only, it is unnecessary to insert the pull down resistor.
- (2) In the case of using V<sub>OT</sub> with low voltage controller like 3.3V MCU, V<sub>OT</sub> output might exceed control supply voltage 3.3V when temperature rises excessively. If system uses low voltage controller, it is recommended to insert a clamp Di between control supply of the controller and V<sub>OT</sub> output for preventing over voltage destruction.
- (3) In the case of not using  $V_{OT}$ , leave  $V_{OT}$  output NC (No Connection).

Refer the application note for this product about the usage of Vor.

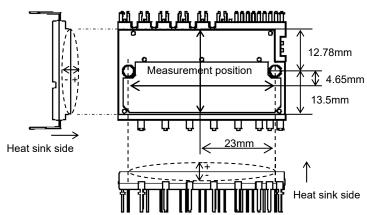
# TRANSFER MOLDING TYPE INSULATED TYPE

#### **MECHANICAL CHARACTERISTICS AND RATINGS**

Parameter	Condition			Limits		
Parameter				Тур.	Max.	Unit
Mounting torque	Mounting screw : M3 (Note 7) Recommended 0.78N·m			-	0.98	N·m
Terminal pulling strength	Load 9.8N JEITA-ED-4701		10	-	-	s
Terminal bending strength	Load 4.9N, 90deg. bend JEITA-ED-4701			-	-	times
Weight				21	-	g
Heat-sink flatness	(Note 8)			-	100	μm

Note 7: Plain washers (ISO 7089~7094) are recommended.

Note 8: Measurement point of heat sink flatness



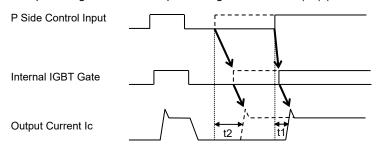
#### **RECOMMENDED OPERATION CONDITIONS**

Symbol	Parameter	Con	dition		Limits			Unit
Symbol	Farameter	Con	dition		Min.	Тур.	Max.	Offic
V <sub>CC</sub>	Supply voltage	Applied between P-NU, NV, NW			0	300	400	V
V <sub>D</sub>	Control supply voltage	Applied between V <sub>P1</sub> -V <sub>NC</sub> , V	/ <sub>N1</sub> -V <sub>NC</sub>		13.5	15.0	16.5	V
$V_{DB}$	Control supply voltage	Applied between V <sub>UFB</sub> -V <sub>UFS</sub>	, V <sub>VFB</sub> -V <sub>VFS</sub> , V	wfb-Vwfs	13.0	15.0	18.5	V
$\Delta V_D$ , $\Delta V_{DB}$	Control supply variation				-1	-	+1	V/µs
t <sub>dead</sub>	Arm shoot-through blocking time	For each input signal		1.5	-	-	μs	
f <sub>PWM</sub>	PWM input frequency	T <sub>C</sub> ≤ 100°C, T <sub>i</sub> ≤ 125°C		-	-	20	kHz	
	Allerentelle man en annual f	$V_{CC} = 300V, V_D = 15V, P.F = 0.8,$		f <sub>PWM</sub> = 5kHz	-	-	14.0	
I <sub>O</sub>	Allowable r.m.s. current	Sinusoidal PWM T <sub>C</sub> ≤ 100°C, T <sub>i</sub> ≤ 125°C	(Note9)	f <sub>PWM</sub> = 15kHz	-	-	13.0	Arms
PWIN(on)				(Note 10)	0.7	-	-	
	1	200V≤V <sub>CC</sub> ≤350V,	Below rate	d current	1.4		-	
PWIN(off)	Minimum input pulse width	13.5V≤V <sub>D</sub> ≤16.5V, 13.0V≤V <sub>DB</sub> ≤18.5V, -20°C≤Tc≤100°C.	Between rated current and 1.7 times of rated current  Between 1.7 times and 2.0 times of rated current		2.5	-	-	μs
		N-line wiring inductance less than 10nH (Note 11)			3.0	-	-	
V <sub>NC</sub>	V <sub>NC</sub> variation	Between V <sub>NC</sub> -NU, NV, NW (including surge)		-5.0	-	+5.0	V	
Tj	Junction temperature				-20	-	+125	°C

Note 9: Allowable r.m.s. current depends on the actual application conditions.

10: DIPIPM might not make response if the input signal pulse width is less than PWIN(on)

#### Delayed Response against Shorter Input Off Signal than PWIN(off) (P-side only)



Real line: off pulse width > PWIN(off); turn on time t1 Broken line: off pulse width < PWIN(off); turn on time t2 (t1:Normal switching time)

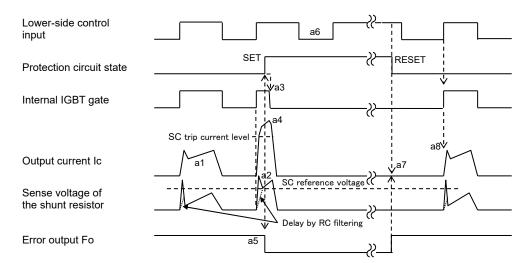
<sup>11:</sup> IPM might make delayed response or no response for the input signal with off pulse width less than PWIN(off). Please refer below about delayed response.

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#### Fig. 5 Timing Charts of The DIPIPM Protective Functions

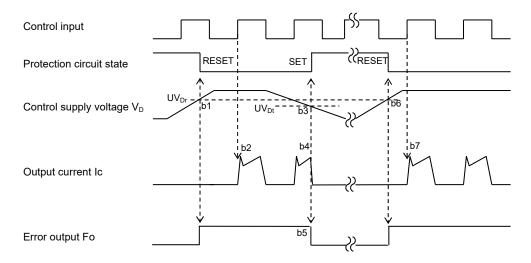
#### [A] Short-Circuit Protection (N-side only with the external shunt resistor and RC filter)

- a1. Normal operation: IGBT ON and outputs current.
- a2. Short circuit current detection (SC trigger)
  - (It is recommended to set RC time constant 1.5~2.0µs so that IGBT shut down within 2.0µs when SC.)
- a3. All N-side IGBT's gates are hard interrupted.
- a4. All N-side IGBTs turn OFF.
- a5.  $F_0$  outputs. The pulse width of the Fo signal is set by the external capacitor  $C_{FO}$ .
- a6. Input = "L": IGBT OFF
- a7. Fo finishes output, but IGBTs don't turn on until inputting next ON signal (L $\rightarrow$ H).
- (IGBT of each phase can return to normal state by inputting ON signal to each phase.)
- a8. Normal operation: IGBT ON and outputs current.



#### [B] Under-Voltage Protection (N-side, UV<sub>D</sub>)

- b1. Control supply voltage V<sub>D</sub> exceeds under voltage reset level (UV<sub>Dr</sub>), but IGBT turns ON by next ON signal (L→H). (IGBT of each phase can return to normal state by inputting ON signal to each phase.)
- b2. Normal operation: IGBT ON and outputs current.
- b3.  $V_D$  level drops to under voltage trip level. (UV $_{Dt}$ ).
- b4. All N-side IGBTs turn OFF in spite of control input condition.
- b5. Fo outputs for the period set by the capacitance  $C_{FO_i}$  but output is extended during  $V_D$  keeps below  $UV_{Dr}$ .
- b6. V<sub>D</sub> level reaches UV<sub>Dr</sub>.
- b7. Normal operation: IGBT ON and outputs current.



Publication Date: January 2021

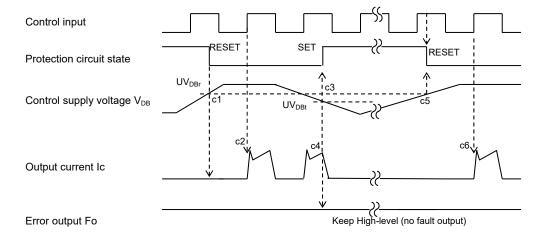
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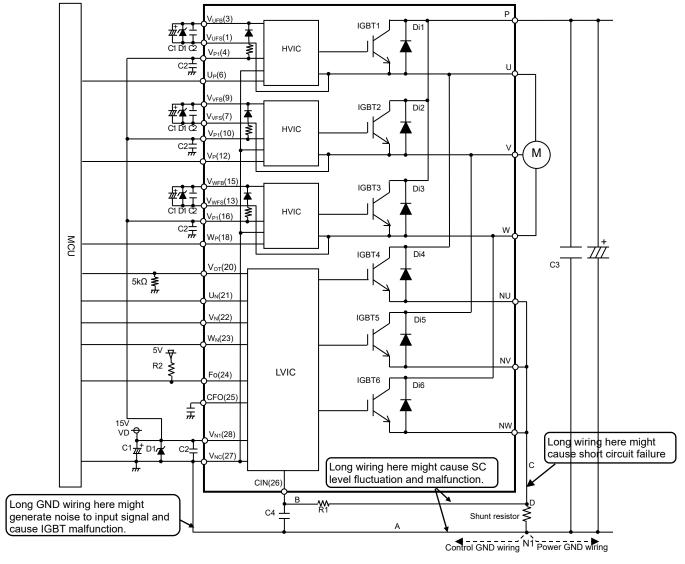
#### [C] Under-Voltage Protection (P-side, UV<sub>DB</sub>)

- c1. Control supply voltage  $V_{DB}$  rises. After the voltage reaches under voltage reset level  $UV_{DBr}$ , IGBT turns on by next ON signal (L $\rightarrow$ H).
- c2. Normal operation: IGBT ON and outputs current.
- c3. V<sub>DB</sub> level drops to under voltage trip level (UV<sub>DBt</sub>).
- c4. IGBT of the correspond phase only turns OFF in spite of control input signal level, but there is no Fo signal output.
- c5.  $V_{DB}$  level reaches  $UV_{DBr}$ .
- c6. Normal operation: IGBT ON and outputs current.



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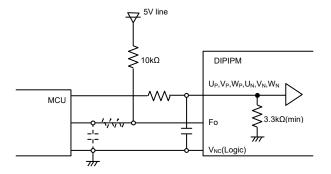
Fig. 6 Example of Application Circuit



- (1) If control GND is connected with power GND by common broad pattern, it may cause malfunction by power GND fluctuation. It is recommended to connect control GND and power GND at only a point N1 (near the terminal of shunt resistor).
- (2) It is recommended to insert a Zener diode D1(24V/1W) between each pair of control supply terminals to prevent surge destruction.
- (3) To prevent surge destruction, the wiring between the smoothing capacitor and the P, N1 terminals should be as short as possible. Generally a 0.1-0.22µF snubber capacitor C3 between the P-N1 terminals is recommended.
- (4) R1, C4 of RC filter for preventing protection circuit malfunction is recommended to select tight tolerance, temp-compensated type. The time constant R1C4 should be set so that SC current is shut down within 2µs. (1.5µs~2µs is recommended generally.) SC interrupting time might vary with the wiring pattern, so the enough evaluation on the real system is necessary.
- (5) To prevent malfunction, the wiring of A, B, C should be as short as possible.
- (6) The point D at which the wiring to CIN filter is divided should be near the terminal of shunt resistor. NU, NV, NW terminals should be connected at near NU, NV, NW terminals when it is used by one shunt operation. Low inductance SMD type with tight tolerance, temp-compensated type is recommended for shunt resistor.
- (7) All capacitors should be mounted as close to the terminals as possible. (C1: good temperature, frequency characteristic electrolytic type and C2:0.22μ-2μF, good temperature, frequency and DC bias characteristic ceramic type are recommended.)
- (8) Input logic is High-active. There is a 3.3kΩ(min.) pull-down resistor in the input circuit of IC. To prevent malfunction, the input wiring should be as short as possible. When using RC coupling, make the input signal level meet the turn-on and turn-off threshold voltage.
- (9) Fo output is open drain type. It should be pulled up to power supply of MCU (e.g. 5V,3.3V) by a resistor that makes I<sub>Fo</sub> up to 1mA. (I<sub>FO</sub> is estimated roughly by the formula of control power supply voltage divided by pull-up resistance. In the case of pulled up to 5V, 10kΩ (5kΩ or more) is recommended.) When using opto coupler, Fo also can be pulled up to 15V (control supply of DIPIPM) by the resistor.
- (10) Fo pulse width can be set by the capacitor connected to CFO terminal.  $C_{FO}(F) = 9.1 \times 10^{-6} \times t_{FO}$  (Required Fo pulse width).
- (11) If high frequency noise superimposed to the control supply line, IC malfunction might happen and cause DIPIPM erroneous operation. To avoid such problem, line ripple voltage should meet dV/dt ≤+/-1V/μs, Vripple≤2Vp-p.
- (12) For DIPIPM, it isn't recommended to drive same load by parallel connection with other phase IGBT or other DIPIPM.

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Fig. 7 MCU I/O Interface Circuit

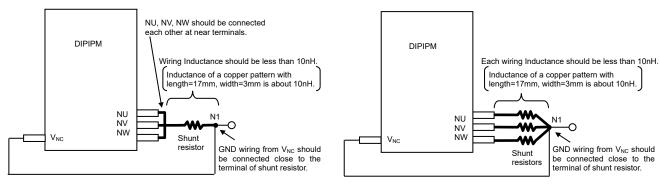


#### Note)

Design for input RC filter depends on PWM control scheme used in the application and wiring impedance of the printed circuit board. DIPIPM input signal interface integrates a minimum  $3.3 k\Omega$  pull-down resistor. Therefore, when inserting RC filter, it is necessary to satisfy turn-on threshold voltage requirement.

Fo output is open drain type. It should be pulled up to control power supply (e.g. 5V, 15V) with a resistor that makes Fo sink current  $I_{Fo}$  1mA or less. In the case of pulled up to 5V supply,  $10k\Omega$  ( $5k\Omega$  or more) is recommended.

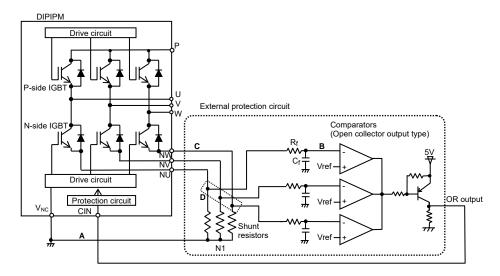
Fig. 8 Pattern Wiring Around the Shunt Resistor



Low inductance shunt resistor like surface mounted (SMD) type is recommended.

#### Fig. 9 Pattern Wiring Around the Shunt Resistor (for the case of open emitter)

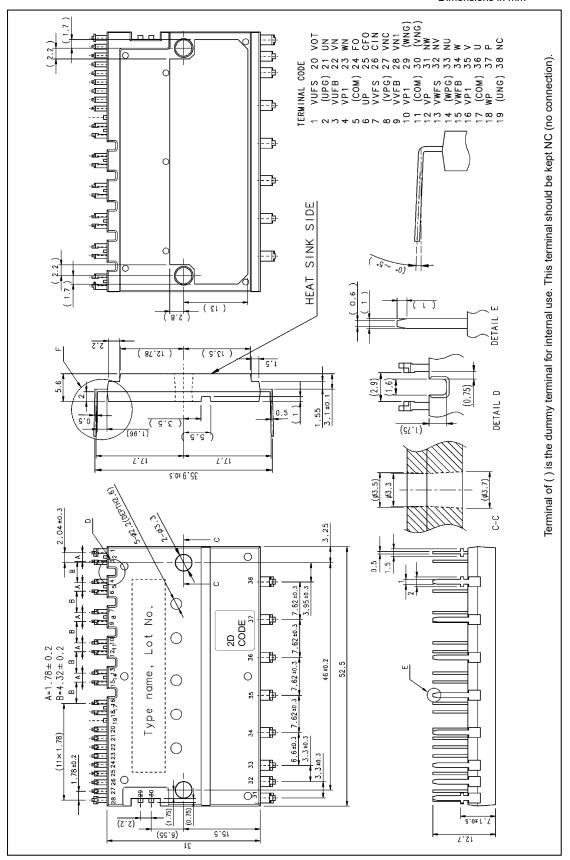
When DIPIPM is operated with three shunt resistors, voltage of each shunt resistor cannot be input to CIN terminal directly. In that case, it is necessary to use the external protection circuit as below.



- It is necessary to set the time constant R<sub>f</sub>C<sub>f</sub> of external comparator input so that IGBT stops within 2μs when short circuit occurs.
   SC interrupting time might vary with the wiring pattern, comparator speed and so on.
- (2) It is recommended for the threshold voltage Vref to set to the same rating of short circuit trip level (Vsc(ref): typ. 0.48V).
- (3) Select the external shunt resistance so that SC trip-level is less than specified value (=2.0 times of rating current).
- (4) To avoid malfunction, the wiring A, B, C should be as short as possible.
- (5) The point D at which the wiring to comparator is divided should be close to the terminal of shunt resistor.
- (6) OR output high level when protection works should be over 0.51V (=maximum Vsc(ref) rating).
- (7) GND of Comparator, GND of Vref circuit and Cf should be not connected to power GND but to control GND wiring.

Fig. 10 Package Outlines

Dimensions in mm



## **Important Notice**

The information contained in this datasheet shall in no event be regarded as a guarantee of conditions or characteristics. This product has to be used within its specified maximum ratings, and is subject to customer's compliance with any applicable legal requirement, norms and standards.

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Publication Date : January 2021 MITSUBISHI ELECTRIC CORPORATION

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